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# Feasibility and parameter optimization of ground-to-satellite uplink continuous-variable quantum key distribution

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## Abstract

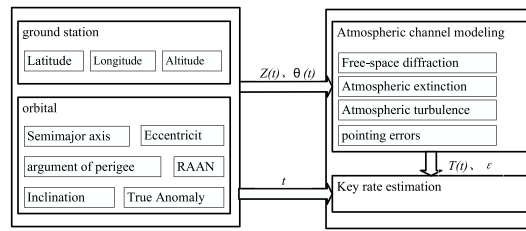
Driven by the necessities of wide-area quantum secure communication networks, ground-to-satellite quantum key distribution (QKD) has been highlighted in the field of quantum information. Continuous variable QKD (CV-QKD) shows the advantages of a high secure key rate generation and compatibility in the fiber channel, but the communication range is limited by low attenuation tolerance. In this study, the feasibility of the uplink CV-QKD for Very Low Earth Orbit satellite is studied, and the influence of practical factors such as atmospheric turbulence and beam pointing errors under dynamic time-varying satellite orbital parameters is analyzed to model realistic transmission channels. A parameter optimization scheme for orbital parameters, spot size, receiving telescope aperture, and modulation variance has been proposed to enhance the performance of the actual uplink CV-QKD, providing a reference for parameter selection in CV-QKD experiments.

## 1. Introduction

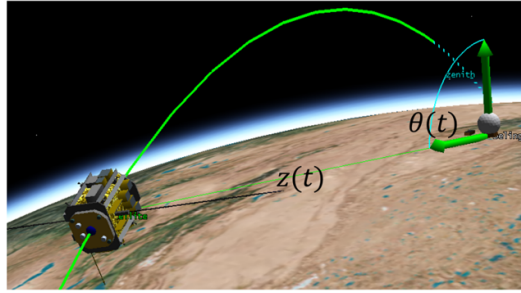
Quantum key distribution (QKD) provides secure keys guaranteed by quantum mechanics [1]. At present, the fiber-based QKD protocols are relatively mature [2], and the quantum satellite ‘Micius’ verifies the feasibility of satellite-to-ground quantum teleportation, discrete variable QKD (DV-QKD), and entanglement distribution experiments [3], which validated the feasibility of free space DV-QKD, making the wide-area QKD network possible [4].

Compared with the DV-QKD, the CV-QKD has the advantages of higher secure rate and compatibility in optical fiber channels [5], which makes it likely to be generalized to space channel. Although the research work on the feasibility of the CV-QKD in space is in the early stages of theoretical research, it is still necessary to analyze the feasibility of the protocol in the actual environment. Dequal *et al* [6] analyzed the downlink Gaussian modulation CV-QKD in low-orbit satellites under reverse reconciliation. References [7, 8] analyzed the theoretical limit and the practical security rate of coherent state transmission in turbulent channels. In [9], the noiseless attenuation scheme is assumed to analyze the performance of uplink CV-QKD. The previous studies have adopted ideal zenith-crossing circular orbit models with fixed-parameter. This paper delves into the feasibility and parameter optimization of ground-to-satellite uplink CV-QKD. It employs a dynamic operational orbit model that reflects real conditions of satellites by incorporating actual orbit parameters and accounting for the dynamic movements of satellites around the Earth. This model is more significantly affected by turbulence [10].

In this study, the feasibility of Gaussian modulation CV-QKD in uplink is investigated by simulation based on the time-varying orbital motion model. The influence of practical factors such as diffraction, extinction, turbulence, and pointing errors on the uplink channel is analyzed to model realistic transmission channels. A parameter optimization scheme for orbital parameters, spot size, receiving telescope aperture, and modulation variance has been proposed to enhance the performance of the actual uplink CV-QKD, which provides theoretical support for subsequent space channel CV-QKD experiments.



**Figure 1.** The framework for modeling uplink CV-QKD.



**Figure 2.** Schematic diagram of CV-QKD communication scheme. A fixed ground station (Alice) and a satellite (Bob) follows a self-defined trajectory.

## 2. Protocol model

This study constructed a model of the ground-to-satellite uplink CV-QKD, which includes satellite orbit modeling, atmospheric channel modeling, and key rate estimation. The framework for the modeling shown in figure 1 is described as follows.

*Step 1.* Longitude, latitude, and altitude are utilized to establish the ground station's geographic coordinates. The semimajor axis, eccentricity, argument of perigee, RAAN, inclination, and true anomaly are parameters integral to the construction of the satellite's orbital model. The foundational architecture of the scenarios is demonstrated by the integration of ground stations and the satellite orbit, which provides parameters such as access duration and link geometry information to determine the time  $t$ , zenith angle  $\theta(t)$  and distance  $z(t)$  between the satellite and the ground station.

*Step 2.* In the atmospheric channel model, the link loss is calculated by analyzing factors such as free-space diffraction, atmospheric extinction, atmospheric turbulence, and pointing errors, all predicated on dynamic parameters.

*Step 3.* At last, the secret key rate can be estimated to analyze the feasibility of the CV-QKD.

### 2.1. Communication modeling

In this study, the detection system is situated on the satellite, while the coherent source is positioned at the ground station. The satellite orbits at an altitude  $h$  beyond the Kármán line ( $h \geq 100\text{km}$ ). The zenith angle  $\theta(t)$  and the slant distance  $z(t)$  between the satellite and the ground station are time-dependent functions. The geometric relationship is illustrated in figure 2. This research employs the Satellite Toolkit to construct self-defined satellite and ground station models [11].

At the ground station, we have a tracking system in place to follow the satellite's trajectory, a pilot signal generator that transmits reference symbols, and an adaptive optics (AO) system [12–15] which is worked in a pre-compensation regime. The satellite receiver is equipped with a pilot signal generator. The implementation of the transmit AOs requires Bob to send a strong beacon light to Alice before the communication officially begins. Alice then uses the wavefront phase information of the beacon light to pre-correct the wavefront of the quantum signal being sent [16].

The satellite receiver is equipped with a fiber-coupled coherent detector with a free-running local oscillator (local LO).

### 2.2. Atmospheric channel modeling

During the beam propagation process, various disturbance effects come into play [17], with the severity being intimately linked to the satellite's running time  $t$ . In this study, the effects of free-space diffraction,

atmospheric extinction, weak atmospheric turbulence and pointing errors on uplink beam propagation at night are analyzed to calculate the transmission efficiency.

In the context of weak atmospheric turbulence at night, and with the incorporation of an AOs system to correct for turbulence, subsequent simulation analyses will exclude the probability of interruptions caused by fluctuations in the angle of arrival [16].

### 2.2.1. Free-space diffraction

We postulate that the ground station emits a collimated Gaussian beam characterized by a carrier wavelength of  $\lambda = 800\text{nm}$ . The beam's initial spot size is denoted as  $w_0$ , and the radius of curvature [18] is  $R_0 = \infty$ . When the beam has propagated a distance  $z(t)$ , the spot size at the receiver can be expressed by the Gaussian beam propagation equation [7]:

$$w_d(t) = w_0 \sqrt{\left(1 - \frac{z(t)}{R_0}\right)^2 + \left(\frac{z(t)}{z_R}\right)^2} \quad (1)$$

where  $z_R = \pi w_0^2 \lambda^{-1}$  is the Rayleigh range. Since the radius of the receiver telescope on the satellite is  $a_R$ , the diffraction-induced transmission efficiency [7] can be given by:

$$\eta_d(t) = 1 - e^{-\frac{2a_R^2}{w_d^2(t)}}. \quad (2)$$

### 2.2.2. Atmospheric extinction

Considering the zenith angle  $\theta(t)$  between the satellite and the ground station, the atmospheric transmission efficiency [6] can be approximated as:

$$\eta_{\text{atm}}(t) \approx \eta_{\text{atm}}^{\text{zen}} \sec(\theta(t)). \quad (3)$$

Where  $\eta_{\text{atm}}^{\text{zen}} = e^{-\int_0^h dh' \alpha(h')} \geq e^{-\alpha_0 \tilde{h}} \approx 0.967$  [8] is the transmission efficiency at the zenith.  $\alpha(h) = \alpha_0 e^{-h/\tilde{h}}$  is the extinction factor, where  $\tilde{h} = 6600\text{ m}$ .  $\alpha_0 = 5 \times 10^{-6} \text{ m}^{-1}$  at  $\lambda = 800\text{nm}$ .

### 2.2.3. Atmospheric turbulence

Atmospheric turbulence can induce beam deflection and broadening, which in turn leads to signal fading. Specifically, in the uplink, the turbulence causes the spot size on the receiver to increase according to the following relationship:

$$w_{\text{st}}^2(t) \approx w_d^2(t) + z^2 \Delta(\theta(t)) \quad (4)$$

where  $\Delta(\theta(t)) = \frac{26.28 I_\infty \sec^6 \theta^{6/5}}{\lambda^{2/5}} - \frac{7.71 I_\infty \sec \theta}{w_0^{1/3}}$ . In these formulas,  $I_\infty$  takes different values depending on the atmospheric model (H-V model). This manuscript takes  $I_\infty \approx 2.2354 \times 10^{-12} \text{ m}^{1/3}$  at night, and  $I_\infty \approx 23.2854 \times 10^{-12} \text{ m}^{1/3}$  at day [8].

Therefore, diffraction-induced transmission efficiency is replaced by turbulence-induced transmission efficiency [7]:

$$\eta_{\text{st}}(t) = 1 - e^{-\frac{2a_R^2}{w_{\text{st}}^2(t)}}. \quad (5)$$

Considering the free-space diffraction, atmospheric extinction, atmospheric turbulence and inevitable fixed loss [6]  $\eta_{\text{ATT}}$  (assumes  $\eta_{\text{ATT}} = 3.8\text{ dB}$ , including 3 dB of losses for fiber coupling and an additional 0.8 dB for taking into account the fact that we are only considering the main peak of the Airy diffraction pattern), the total transmission efficiency can be given by:

$$\eta_{\text{tot}}(t) = \eta_{\text{ATT}} \eta_{\text{atm}}(t) \eta_{\text{st}}(t). \quad (6)$$

It is worth noting that equation (6) represents the transmission efficiency without considering the beam deflection. When beam deflection is taken into account, equation (6) denotes the maximum achievable transmission efficiency. The beam deflection is characterized by the wandering of the beam's centroid according to a Gaussian distribution with variance given by [8]:

$$\sigma_{\text{TB}}^2(t) \approx \frac{7.71 I_\infty}{w_0^{1/3}} z^2 \sec \theta(t). \quad (7)$$

### 2.2.4. Pointing errors

Pointing errors can also lead to beam deflect. We assumed  $\theta_p = 10^{-6}$  rad, which leads to another random deflection, whose variance [8] can be approximated as:

$$\sigma_p^2(t) \approx (\theta_p z(t))^2. \quad (8)$$

Therefore, the total variance of the random deflection of the beam's centroid can be given by:

$$\sigma^2(t) = \sigma_p^2(t) + \sigma_{TB}^2(t). \quad (9)$$

The deflection of the beam's centroid will cause the instantaneous transmission efficiency to be  $T(t) \leq \eta_{\text{tot}}(t)$ .

In case of weak turbulence and considering a wavelength of 800 nm, the probability distribution of the deflection distance follows the Wei-bull distribution [19], and  $T(t)$  obeys the probability distribution  $P_\sigma(T)$  [8]:

$$P_\sigma(T) = \frac{r_0^2}{\gamma \sigma^2 T(t)} \left( \ln \frac{\eta_{\text{tot}}(t)}{T(t)} \right)^{\frac{2}{\gamma}-1} \exp \left[ -\frac{r_0^2}{2\sigma^2} \left( \ln \frac{\eta_{\text{tot}}(t)}{T(t)} \right)^{\frac{2}{\gamma}} \right] \quad (10)$$

where

$$\gamma(t) = \frac{4 \frac{2a_R^2}{w_{\text{st}}^2(t)} f_0 \left( \frac{2a_R^2}{w_{\text{st}}^2(t)} \right) f_1 \left( \frac{2a_R^2}{w_{\text{st}}^2(t)} \right)}{\ln \left[ 2\eta_{\text{st}}(t) f_0 \left( \frac{2a_R^2}{w_{\text{st}}^2(t)} \right) \right]} \quad (11)$$

$$r_0(t) = \frac{a_R}{\ln \left[ 2\eta_{\text{st}}(t) f_0 \left( \frac{2a_R^2}{w_{\text{st}}^2(t)} \right) \right]^{1/\gamma(t)}} \quad (12)$$

$$f_0(x) = [1 - \exp(-2x) I_0(2x)]^{-1} \quad (13)$$

$$f_1(x) = \exp(-2x) I_1(2x) \quad (14)$$

$I_n$  is the first kind of modified Bessel function with order  $n = 0, 1$ .

### 2.3. Secure key rate estimation

According to the previous analysis of the atmospheric channel, the actual instantaneous transmission efficiency  $T(t)$  is a random variable. We propose employing a beacon signal to estimate the instantaneous channel transmission efficiency. The fading channel can be modeled as an equivalent non-fading channel, characterized by a fixed effective transmittance  $T_e(t)$ .

Given the relatively small additional noise caused by phase recuperation [6, 8], this paper treats the excess noise as a single parameter, denoted by  $\varepsilon$ , which represents the excess noise at Alice's side.

For the excess noise  $\varepsilon$ ,  $\varepsilon_e(t)$  is used as the equivalent excess noise [16, 20].

$$\begin{cases} T_e(t) = E \left[ \sqrt{T(t)} \right]^2 \\ \varepsilon_e(t) = \frac{E[T(t)]}{E \left[ \sqrt{T(t)} \right]^2} \varepsilon + \frac{\text{Var} \left( \sqrt{T(t)} \right) (V-1)}{E \left[ \sqrt{T(t)} \right]^2} \end{cases} \quad (15)$$

Where  $E \left[ \sqrt{T(t)} \right] = \int_0^{\sqrt{\eta_{\text{tot}}(t)}} \sqrt{T(t)} P \left( \sqrt{T(t)} \right) d\sqrt{T(t)}$  is the expectation of  $\sqrt{T(t)}$ , and  $\text{Var} \left( \sqrt{T(t)} \right) = E[T(t)] - E \left[ \sqrt{T(t)} \right]^2$  describes the strength of the channel fluctuations.

In the case of reverse reconciliation, the asymptotic value of the secret key rate can be written as [20, 21]:

$$R = \beta I_{AB} - \chi_{BE}. \quad (16)$$

The channel-added noise referred to the channel input, is defined as  $\chi_{\text{line}} = \frac{1}{T_e} - 1 + \varepsilon_e$ . An actual detector has the detection efficiency  $\eta$  and the electronic noise vel. The detection-added noise referred to Bob's input is defined as  $\chi_{\text{hom}} = [(1 - \eta) + \text{vel}] / \eta$  and  $\chi_{\text{het}} = [1 + (1 - \eta) + 2\text{vel}] / \eta$  for homodyne and heterodyne detection, respectively. So, the total noise referred to the channel input can then be expressed as  $\chi_{\text{tot}} = \chi_{\text{line}} + \chi_{\text{h}} / T_e$

Alice prepares the coherent states according to a Gaussian distribution with variance  $V_A N_0$ , where  $N_0$  is the shot noise variance. We can define  $V = V_A + 1$ . The mutual information of Alice and Bob for homodyne and heterodyne detection can be written as:

$$I_{AB}^{\text{hom}} = \frac{1}{2} \log_2 \frac{V + \chi_{\text{tot}}}{1 + \chi_{\text{tot}}} \quad (17)$$

$$I_{AB}^{\text{het}} = \log_2 \frac{V + \chi_{\text{tot}}}{1 + \chi_{\text{tot}}} \quad (18)$$

The maximum amount of information that the Eve can potentially extract from Bob is bounded by the Holevo quantit

$$\chi_{BE} = S(\rho_{AB1}) - S(\rho_{AFG}^{mB}) = \sum_{i=1}^2 G\left(\frac{\lambda_i - 1}{2}\right) - \sum_{i=3}^5 G\left(\frac{\lambda_i - 1}{2}\right) \quad (19)$$

where  $G(x) = (x+1) \log_2(x+1) - x \log_2 x$ , and  $\lambda_{1,2}$  are the symplectic eigenvalues of the covariance matrix  $\gamma_{AB1}$  characterizing the state  $\rho_{AB1}$ , and  $\lambda_{3,4,5}$  are the symplectic eigenvalues of the covariance matrix  $\gamma_{AFG}^{mB}$  characterizing the state  $\rho_{AFG}^{mB}$  [22].

The covariance matrix  $\gamma_{AB1}$  is:

$$\gamma_{AB1} = \begin{bmatrix} V \cdot I & \sqrt{T_e(V^2 - 1)} \cdot \sigma_z \\ \sqrt{T_e(V^2 - 1)} \cdot \sigma_z & T_e(V + \chi_{\text{line}}) \cdot I \end{bmatrix} \quad (20)$$

When we consider the uncertainty of the parameter estimation due to the limited statistics, the Finite-size value of the secret key rate can be written as [23]:

$$R_{\text{finite}} = \frac{n}{N} [\beta I_{AB} - \mathcal{X}_{\epsilon_{\text{PE}}} - \Delta(n)]. \quad (21)$$

Given the clock rate  $C = 1$  GHz, there are  $N = C$  symbols exchanged in 1 s. Out of these  $N$  symbols, only  $n = N - m$  symbols are used for establishing the key, with  $m = 0.1 N$  symbols dedicated to parameter estimation.

What needs to be clarified is that, with a sampling rate higher than the atmospheric coherence time (typically around 1 kHz), the ground station transmits pilot signals and beacon lights, and the detection of the beacon signal's intensity at the satellite receiver can achieve parameter estimation of the transmission efficiency for each point on the orbit during the fading process [6].

$\mathcal{X}_{\epsilon_{\text{PE}}}$  represents the information that Eve manages to intercept when Alice's estimation of Bob's channel parameters has an error probability of  $\epsilon_{\text{PE}}$ . This effect is accounted by considering a lower bound on transmission coefficient  $T_{\min}$  to replace transmission efficiency  $T(t)$  and an upper bound of excess noise  $E(\epsilon_{\max}(t))$  to replace the excess noise  $\epsilon$  [16],

$$\begin{cases} T_{\min}(t) = (\sqcup_{\min}(t))^2 \\ \epsilon_{\max}(t) = \frac{\sigma_{\max}^2(t) - 1}{T_{\min}(t)} \end{cases} \quad (22)$$

where  $z_{\epsilon_{\text{PE}}/2}$  is a parameter related to the estimated probability of failure  $\epsilon_{\text{PE}}$ . (Here we consider  $\epsilon_{\text{PE}} = 10^{-10}$ , which gives  $z_{\epsilon_{\text{PE}}/2} = 6.5$ )

$$\begin{cases} \sqcup_{\min}(t) \approx \sqrt{T(t)} - z_{\epsilon_{\text{PE}}/2} \sqrt{\frac{1+T(t)\epsilon}{mV_A}} \\ \sigma_{\max}^2(t) \approx 1 + T(t)\epsilon + z_{\epsilon_{\text{PE}}/2} \frac{(1+T(t)\epsilon)\sqrt{2}}{\sqrt{m}} \end{cases} \quad (23)$$

The parameter  $\Delta(n)$  is related to the security of the privacy amplification:

$$\Delta(n) \equiv (2\dim\mathcal{H}_x + 3) \sqrt{\frac{\log_2(2/\epsilon)}{n}} + \frac{2}{n} \log_2(1/\epsilon_{\text{PA}}) \quad (24)$$

where  $\mathcal{H}_x$  is the Hilbert space corresponding to the variable  $x$  used in the raw key,  $\epsilon$  is a smoothing parameter, and  $\epsilon_{\text{PA}}$  is the failure probability of the privacy amplification procedure. (Here we consider  $\epsilon = 10^{-10}$ ,  $\epsilon_{\text{PA}} = 10^{-10}$ , and  $\dim\mathcal{H}_x = 2$ .)

The total secure key  $K$  generated in the access duration  $[t_{\text{start}}, t_{\text{end}}]$  can be calculated to be [24]:

$$K = \int_{t_{\text{start}}}^{t_{\text{end}}} R_{\text{finite}}(t) C dt. \quad (25)$$

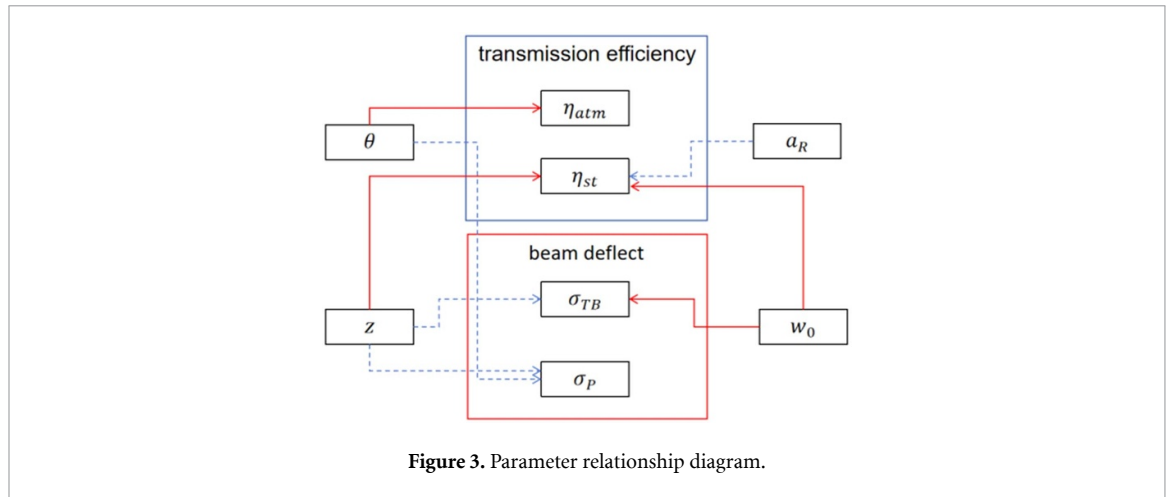


Figure 3. Parameter relationship diagram.

Table 1. Main simulation parameters used in our mode.

Parameter	Symbol	Value
Pointing errors	$\theta_p$	1 $\mu$ rad
Extinction (at 0 rad)	$\eta_{atm}^{zen}$	0.967
Electronic noise	$vel$	0.1 S.N.U.(shot noise unit)
Detector efficiency	$\eta$	0.4
Excess noise	$\varepsilon$	0.01 S.N.U.
Rec. efficiency	$\beta$	0.95
Fixed attenuation	$\eta_{ATT}$	0.4

### 3. Simulation results and parameter analysis

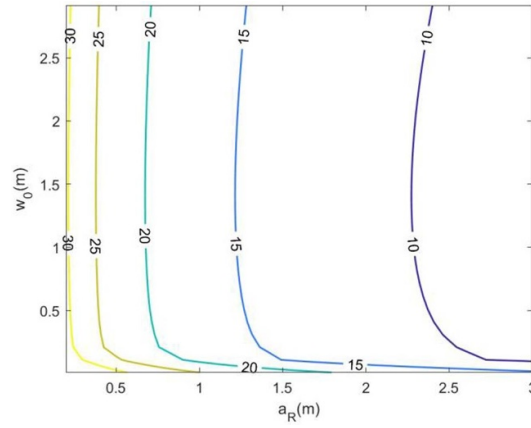
The feasibility of uplink CV-QKD, employing reverse reconciliation and heterodyne detection, is assessed within the aforementioned framework. Atmospheric turbulence exerts a relatively more pronounced influence on the uplink, with the induced beam broadening and deflection being correlated to a range of parameters, including propagation distance, zenith angle, initial spot size, and the aperture of the receiving telescope. As depicted in figure 3, the blue dashed line signifies a positive correlation between the two connected variables, while the red solid line denotes a negative correlation.

The main simulation parameters used in model are shown in table 1.

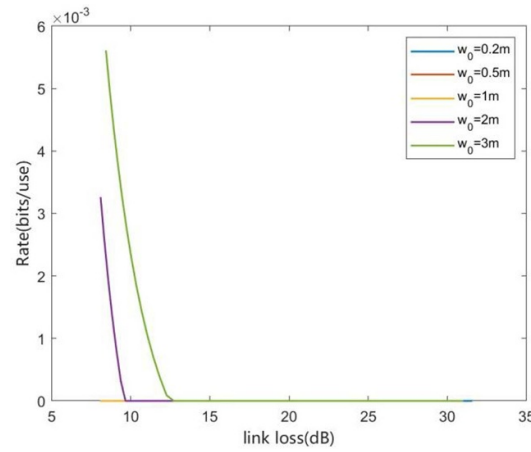
Utilizing real-world parameters as input, our proposed framework serves as a method for conducting a feasibility assessment of space-based CV-QKD within a specific scenario. In this study, we reference the Weber telescope, the largest space optical telescope with a diameter of 6.5 m [25], to limit the parameter boundaries for the receiving telescope aperture and the initial spot size, which are set at 3 m [26, 27]. In this section, we initially conduct a comprehensive exploration of the entire parameter space to find the specific conditions under which feasibility may be achieved. An uplink from the Nagari station to the Micius satellite is modeled to conduct a preliminary feasibility assessment. We conduct an exploration within the parameter space that is compatible with the requirements of the space industry. At the nearest point between the ‘Micius’ and the ground station, (the distance is around 501 km and the satellite is at an zenith angle of 14.1° [11]), The variation of link loss is presented in figure 4. The maximum tolerable loss of CV-QKD is presented in figure 5.

As can be seen from figure 5, the secret key cannot be generated when  $w_0 \leq 1$  m. When  $w_0 = 2$  m, the maximum tolerable attenuation is approaching 10 dB. As illustrated in figure 4, to maintain the link loss beneath 10 dB, a telescope radius  $a_R$  of approximately 2.4 m is necessary. An increased telescope radius can substantially decrease the link loss; however, the size of telescopes that can be feasibly mounted on satellites is constrained by cost and complexity, as larger ones are both more expensive and complicated to implement. In the actual application, a trade-off is inevitable between the secret key rate and the associated costs. The selection of values such as  $w_0 = 2$  m and  $a_R = 2.4$  m is impractical, given the considerable financial outlay.

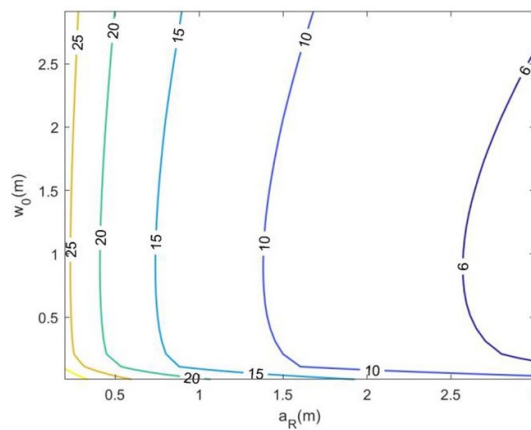
Consequently, the implementation of uplink QKD at the orbit of the ‘Micius’ satellite presents an exceedingly challenge at present. Subsequently, we intend to conduct a parameter analysis at lower orbits to find more viable parameter configurations.



**Figure 4.** Different factors induced link loss (in dB).  $z = 501$  km,  $\theta = 14.1^\circ$ .



**Figure 5.** The tolerable loss of CV-QKD.  $z = 501$  km,  $\theta = 14.1^\circ$ ,  $V = 2$  S.N.U.

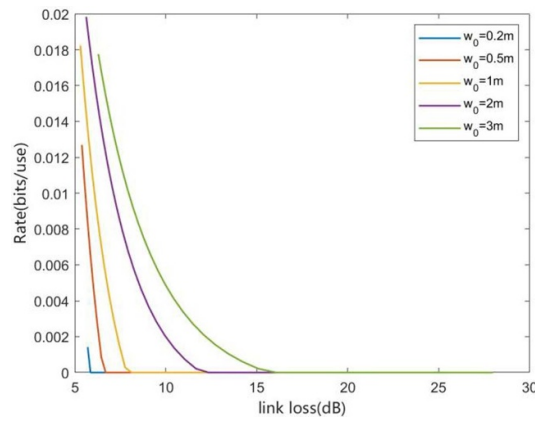


**Figure 6.** Different factors induced link loss (in dB).  $z = 300$  km,  $\theta = 0^\circ$ .

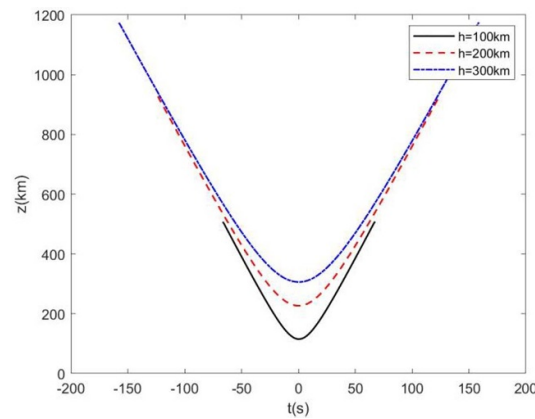
Figures 6 and 7 depict the link loss and the tolerable threshold of link loss for CV-QKD at the zenith of a 300 km orbit, impacted by various parameters. At this juncture, it is evident that  $w_0$  can take on more practical values with  $w_0 \leq 1$  m, while the telescope aperture  $a_R$  still necessitates a significant size to effectively facilitate QKD. Therefore, in the subsequent analysis, we will consider orbital altitudes within 300 km.

Under a certain satellite-based scene model ( $h \leq 300$  km), the influence of three parameters of receiving aperture, beam spot size and modulation variance on the secret key rate is analyzed to explore the parameter





**Figure 7.** The tolerable loss of CV-QKD,  $z = 300$  km,  $\theta = 0^\circ$ ,  $V = 2$  S.N.U.



**Figure 8.** Distance between the satellite and the Delingha station. The lowest distances are approximately 100 km, 200 km and 300 km, respectively.

optimization. It should be pointed out that sats in very low Earth orbits (VLEOs) face a complex dynamic environment and atmospheric resistance, requiring sustained orbital adjustments to mitigate such impacts.

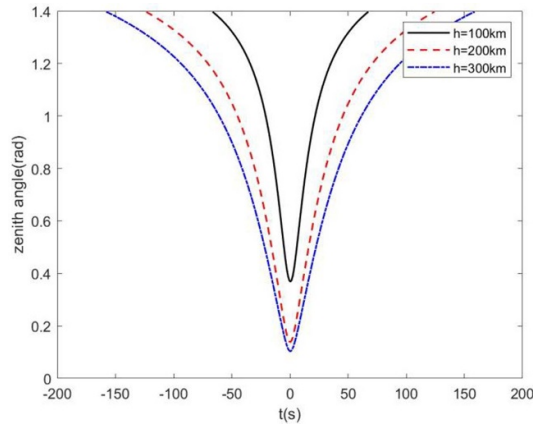
Since most low-orbit satellites have very small ellipticities, mostly in circular or near-circular orbits, this article models dynamic circular orbits with RAAN of  $-17^\circ$ , an inclination of  $55^\circ$ , and different orbital altitude  $h$  of 100 km, 200 km and 300 km. The direct distance  $z(t)$  and the change of the zenith angle  $\theta(t)$  between the satellite and the Delingha station are shown in figures 8 and 9. It should be noted that the satellite orbit is not the zenith-crossing orbit, so the real-time distance  $z(t)$  between the satellite and the ground station is greater than the orbital height  $h$ . The interval is 1 s.

What needs to be pointed out is: due to the favorable assumptions regarding telescope dimensions and orbital parameters that ensure a high transmission rate, we can rely on  $10^8$  symbols to achieve a estimation of parameters and the generation of a finite-length secret key rate.

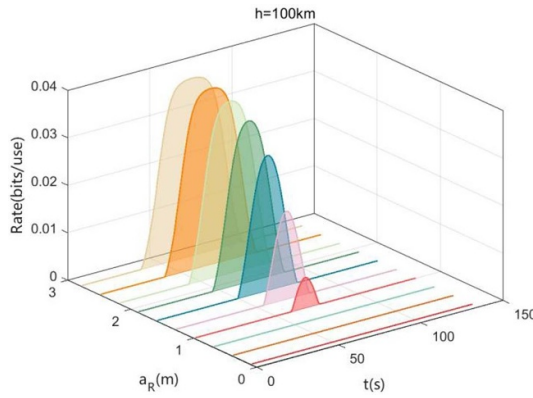
Figures 10 and 11 shows the influence of the aperture size of the receiving telescope on the secret key rate. Figure 10 shows the real-time secure key rate of  $h = 100$  km under different receiving telescope values. It is obvious that the secret key rate and the generation time of the secret key rate both increase with the increasing telescope aperture size. The comparison of the secure key obtained by equation (20) under different receiving telescope values is shown in figure 11.

From figure 11, it is observed that for the utilization of a more practical telescope aperture, with a radius less than 1 m, key rate generation becomes feasible exclusively at an orbital altitude of 100 km. To continue analyzing the impact of other parameters on the key rate at different orbital heights, this paper employs a receiving telescope radius ( $a_R$ ) of 2.4 m in subsequent simulations, a value that enables all three orbits to generate non-zero key rates. To further analyze the influence of various parameters on the key rate across different orbital altitudes, this paper adopts a receiving telescope radius ( $a_R$ ) of 2.4 m for subsequent

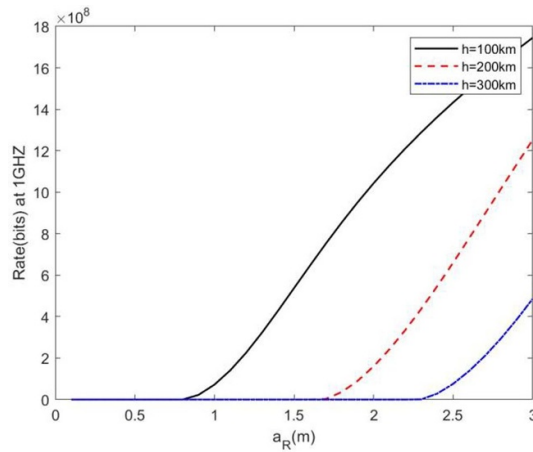




**Figure 9.** Zenith angle between the satellites and the Delingha station. The lowest zenith angles are more than zero.



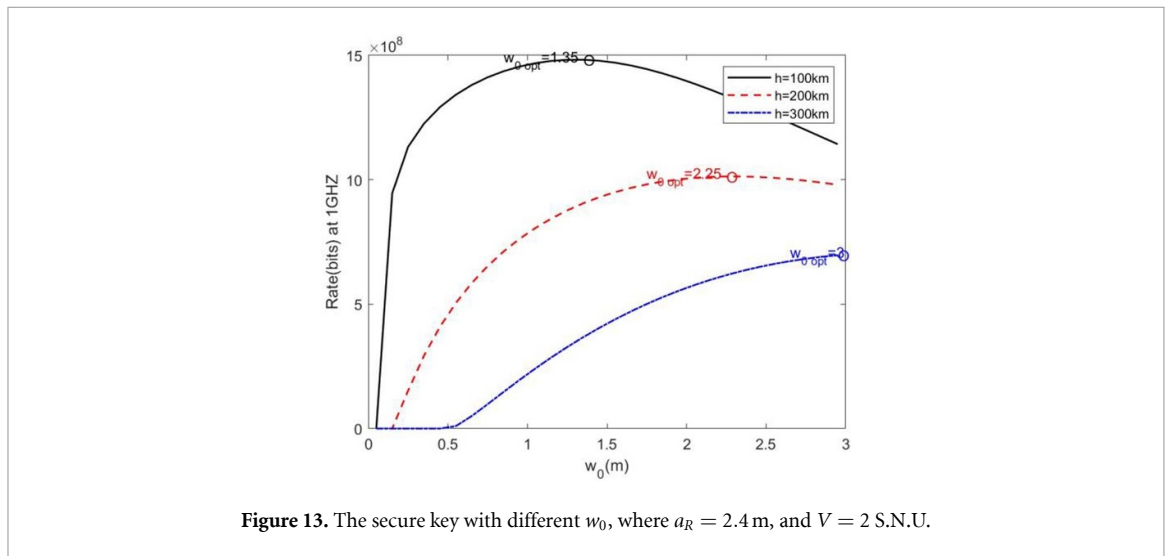
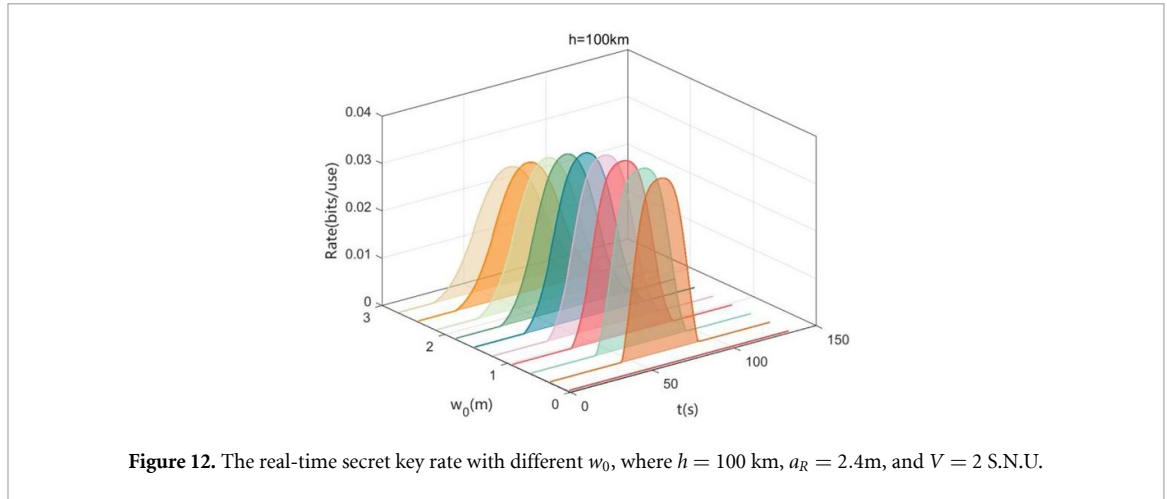
**Figure 10.** The real-time secure key rate under different  $a_R$ , where  $h = 100$  km,  $w_0 = 0.6$  m, and  $V = 2$  S.N.U.



**Figure 11.** The comparison of the secure key under different  $a_R$ , where  $V = 2$  S.N.U, and  $w_0 = 0.6$  m.

simulations. This radius is selected as it facilitates the generation of non-zero key rates for all three considered orbits. However, it must be additionally emphasized that this selected value remains somewhat optimistic and may not fully account for all practical constraints.

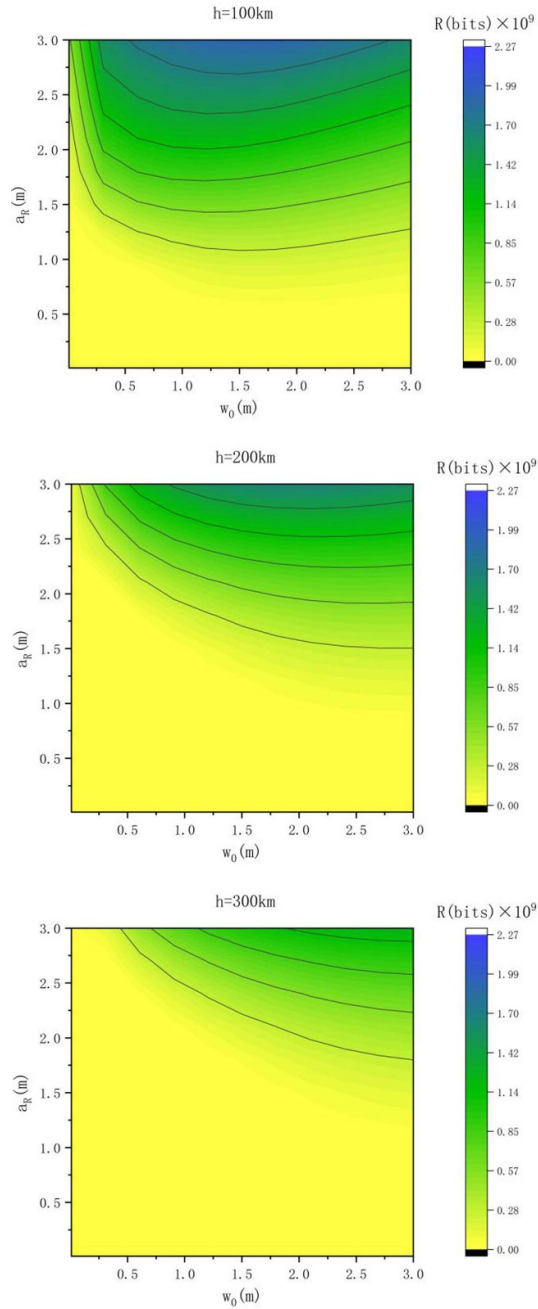
Figures 12 and 13 shows the influence of the initial spot size on the secret key rate ( $a_R = 2.4$  m). Figure 12 present the real-time secure key rate of  $h = 100$  km with different initial spot sizes. It can be seen that there exists an optimal initial spot size to obtain the maximum secret key rate. For the uplink, the transmitter is at



the ground station, and the initial spot size can be adjusted according to the specific satellite orbit to obtain the maximum secret key rate. The secure key generated with different initial spot sizes is shown in figure 13. For the altitude of 100 km, 200 km and 300 km, the optimal  $w_0$  can be set as 1.35 m, 2.25 m, 3 m to obtain the maximized secure key. It is evident that the optimal initial spot size will increase with orbital altitude.

Figure 14 shows the influence of the aperture of the receiving telescope and the initial spot size on the secret key, and (a) (b) (c) are the contour maps of secure key of different altitudes. The blue-green area in the image is a non-zero secret key area, which is concentrated in the upper part of the image. It can be seen that increasing the aperture of the receiving telescope  $a_R$  has a significant effect on improving the secret key rate. By comparing (a) (b) (c), it is visible that the non-zero secret key region shrinks with increasing satellite orbit height. Increasing the aperture of the receiving telescope has a significant effect on improving the secret key rate. The change trend of contour line with  $w_0$ -axis is gentler than that with  $a_R$ -axis, which means  $w_0$  has a relatively small impact on the key rate. Therefore, the importance of the choice of  $a_R$  on the secure key is highlighted.

Figures 15 and 16 shows the impact of modulation variance on secret key rate. The real-time secret key rate of  $h = 100$  km under various modulation variations is illustrated in figure 15. It can be observed that the modulation variance has an effect on the amount of the secret key rate, but has no effect on the duration of the secret key generation. The secret key rate initially increases with the increase of the modulation variance. After the optimal modulation scheme is obtained, the secret key rate decreases with the increase of the modulation variance. Figure 16 reveals the secure key generated under various modulation variances. By comparing the secure key curves at different orbital altitudes, it can be seen that the higher the orbital height,



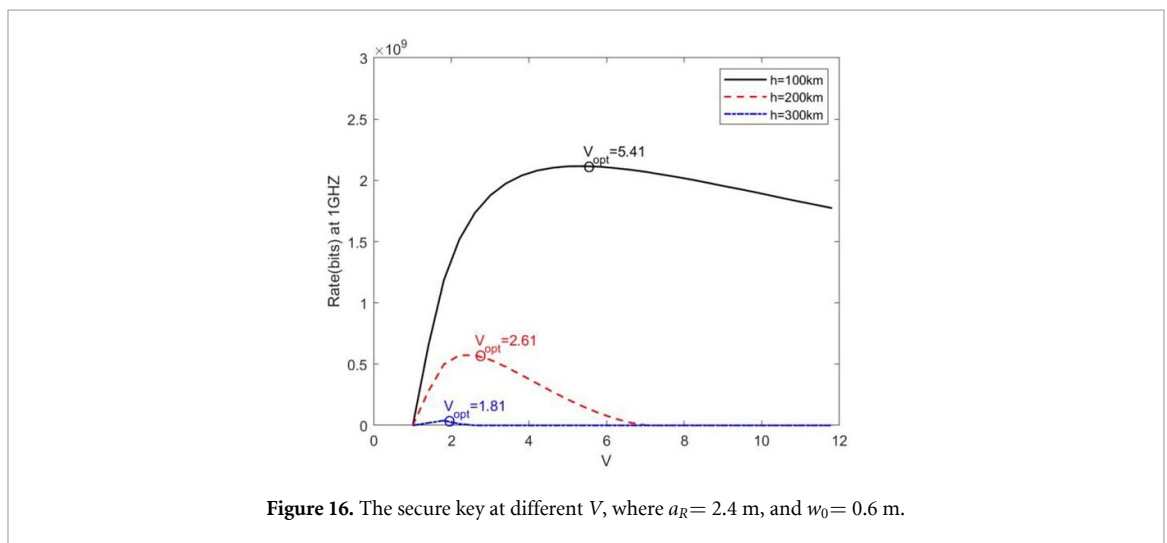
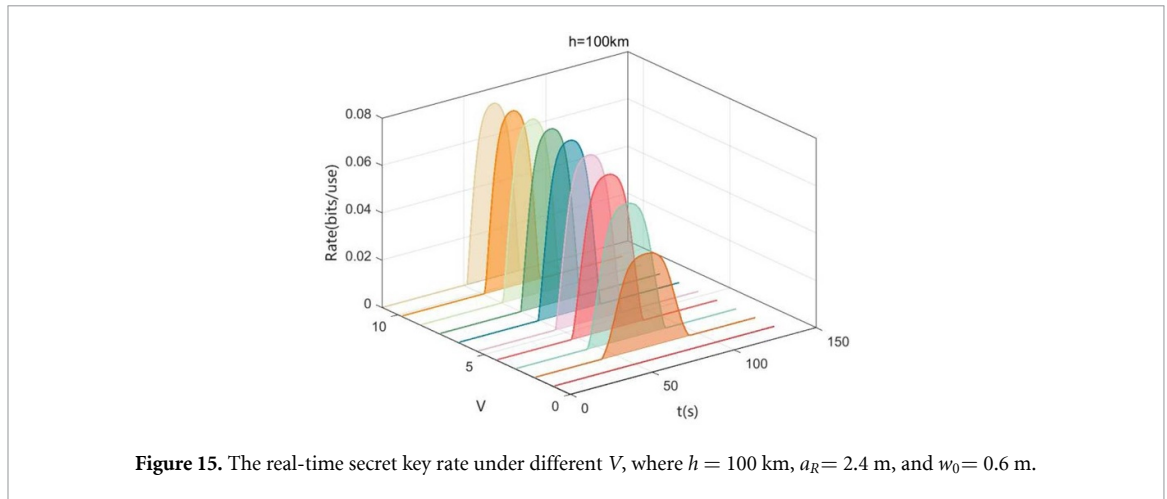
**Figure 14.** The influence of  $w_0$  and  $a_R$  on the secure key, where  $V = 2$  S.N.U. (a)  $h = 100$  km. (b)  $h = 200$  km. (c)  $h = 300$  km.

the smaller the range of the modulation variance that can establish the non-zero secret key. For the altitude of 100 km, 200 km and 300 km, the optimal  $V$  can be set as 5.41, 2.61, 1.81 to obtain the maximized secure key.

After determining the selection of  $a_R$ , the combined effect of the modulation variance and the initial spot size on the secret key rate is shown in figure 17, and (a) (b) (c) are the contour maps of secure key at different altitude. The blue-green region is a non-zero secret key region. In figure 17(a), most regions can generate a secret key, and the dense contour line indicates that a small parameter change can have an obvious impact on the key rate. At the higher orbital altitude  $h$  shown in figure (b)(c), there are fewer optional combinations of  $V_A$  and  $w_0$  that can generate key rates.

Figure 18 shows the secret key rate for different orbital altitudes. Where  $a_R = 2.4$  m, the initial spot size  $w_0$  and the modulation variance  $V_A$  are set to be the optimal value according to the analysis of figure 17.

The orbit height affects the access time and transmission loss of the satellite. The visible communication time between satellite and the ground station is 135 s, 249 s and 318 s, respectively. According to Formula

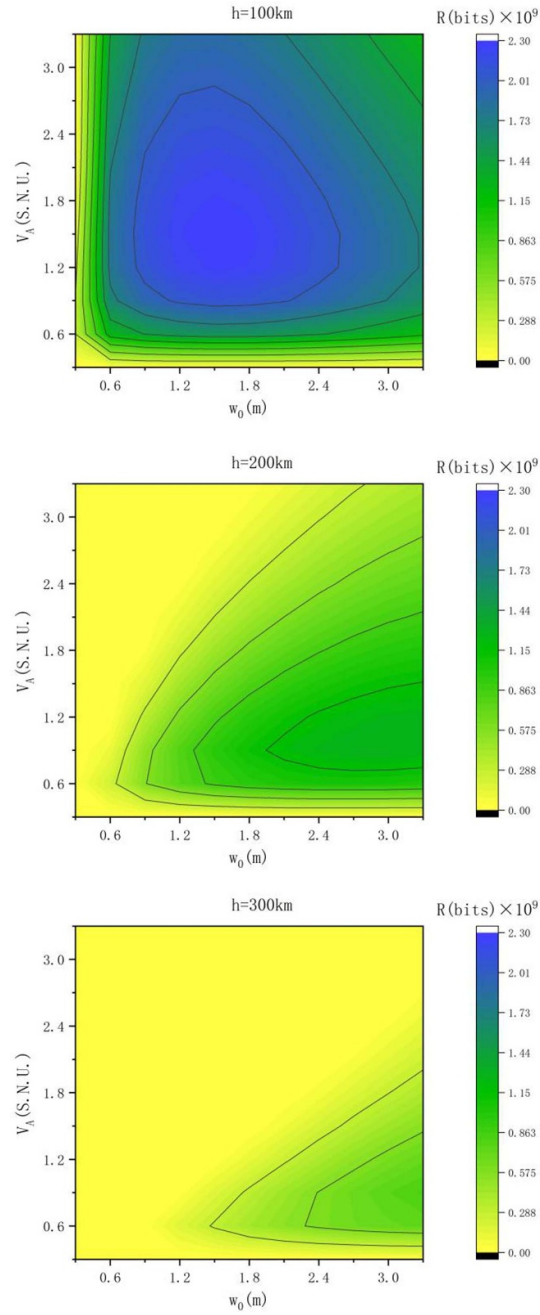


(25), it can be calculated that the secure key of these three satellite orbits during one transmission are  $2.29 \times 10^9$  bits,  $1.29 \times 10^8$  bits, and  $8.23 \times 10^8$  bits, respectively. In this study, the aperture of the receiving telescope is optimistically selected and some optimization parameters are carried out. Compared with the three satellite orbits, the lowest satellite obtains the maximum secure key.

#### 4. Conclusion

In contrast to the traditional fixed circular orbit model, this study adopts a dynamic satellite orbit model for a practical discussion on the feasibility of uplink CV-QKD for VLEO satellites. The research encompasses the modeling of uplink scenarios involving ground station transmission and satellite reception, theoretical analysis of the spatially time-varying degradation channel, and estimation of the key generation rate for the CV-QKD protocol. The feasibility of CV-QKD with dynamic geometric parameter is performed under the uplink scene modeling, which can provide parameters selection and theoretical support for the future free space CV-QKD experiments.

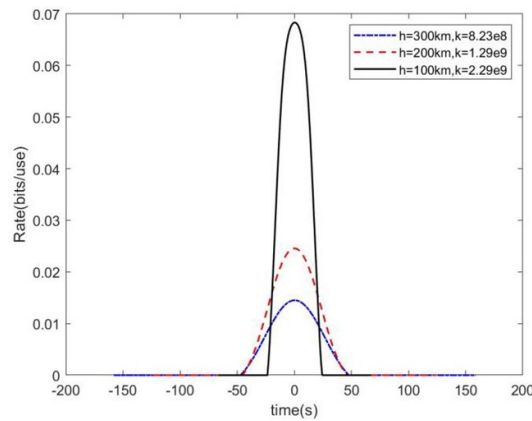
This study also discusses the influence of modulation variance, initial spot size, and receiving telescope aperture on the secret key rate. We demonstrate that the aperture of the receiving telescope and the satellite orbit's altitude, both closely associated with atmospheric turbulence, significantly influence the secret key rate. The simulation outcomes provide optimal parameter selections that yield the highest secure key rates across a range of orbital heights. It is also important to recognize that the parameter selection like  $a_R$  may be overly optimistic. The primary rationale is the challenges encountered in uplink key distribution, notably the non-negligible impact of atmospheric turbulence, an effect that is mitigated in downlink scenarios. Nonetheless, from the analytical standpoint of parameter assessment, this study yields valuable insights and



**Figure 17.** The influence of  $w_0$  and  $V$  on the secret key, where  $a_R = 2.4$  m. (a)  $h = 100$  km. (b)  $h = 200$  km. (c)  $h = 300$  km.

lays a theoretical groundwork, offering guidance for the optimization of parameters in CV-QKD under more realistic conditions in future applications.

Furthermore, subsequent research will focus on harnessing intelligent optimization algorithms for channel compensation, which holds great promise in mitigating the effects of atmospheric turbulence, finite-length effects [23], and the Doppler effect [28]. This approach aims to identify the optimal parameter combinations, thereby enhancing the feasibility and practicality of uplink CV-QKD.



**Figure 18.** The secret key rate for different orbital altitudes where  $a_R = 2.4\text{m}$ , and  $w_0$ ,  $V_A$  are optimized.

## Data availability statement

All data that support the findings of this study are included within the article (and any supplementary files).

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