

# PARAMETER DESIGN OF A ROTATING COIL MEASUREMENT SYSTEM FOR QUADRUPOLES

Y. Xie<sup>†</sup>, H. Liang, W. Chen, J. Yang, B. Qin

State Key Laboratory of Advanced Electromagnetic Engineering and Technology  
College of Electrical and Electronics Engineering  
Huazhong University of Science and Technology, Wuhan, China

## Abstract

This paper describes the design research on a rotating coil measurement system, which is used to measure the integral field harmonics components of quadrupoles in the beamline. The structure of the measurement system, parameters design of the rotating coils and main error analysis are described.

## INTRODUCTION

For the integral harmonic field components measurement of the quadrupole, the rotating coil method has several advantages over Hall probe methods. Hall probe method can't directly measure the harmonic field components. Besides, rotating coil method measures far faster than Hall probe method [1].

The rotating coil measurement system introduced in this paper can be applied to a normal conducting quadrupole with pole aperture 80 mm (L270), which is the main focusing element in a proton therapy beamline. The cross section of the L270 quadrupole are shown in Fig. 1. The main parameters are listed in Table 1.

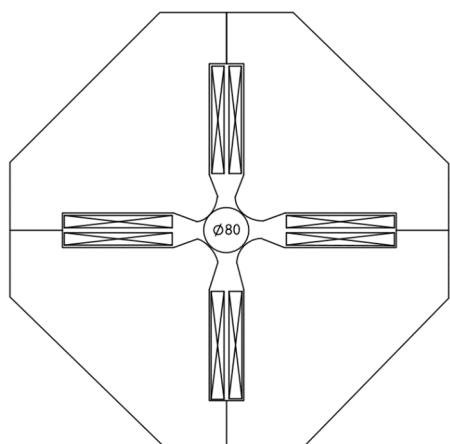


Figure 1: Cross section of the L270 quadrupole.

Table 1: Quadrupole Parameters

Parameters	Value
Pole radius(mm)	40
Radius of good field region(mm)	32
Maximum gradient(T/m)	18.0
Effective length(mm)	270
Magnet yoke length(mm)	240
Higher order harmonic error	$\leq 5.0 \text{E-}4$

<sup>†</sup> yingcai\_xie@hust.edu.cn

## ROTATING COIL PARAMETER DESIGN

### Determination of Radial Size

In Fig. 2, the radius of the primary coil are  $r_1, r_3$ , and the number of turns is  $M_{outer}$ , the radius of the compensated coil are  $r_2, r_4$ , and the number of turns is  $M_{inner}$ .

We define that:

$$\beta_1 = -\frac{r_3}{r_1} = \left| \frac{r_3}{r_1} \right| \quad (1)$$

$$\beta_2 = \left| \frac{r_4}{r_2} \right| \quad (2)$$

$$\rho = \frac{r_2}{r_1} \quad (3)$$

$$\mu = \frac{M_{inner}}{M_{outer}} \quad (4)$$

The coil sensitivity coefficient [3] is expressed as:

$$s_n = 1 - (-\beta_1)^n - \mu \rho^n [1 - (-\beta_2)^n] \quad (5)$$

For a quadrupole (N=2), when high-order quantities are ignored, the expression of the fundamental field [5] is:

$$F = C_2(z^2 + 2z \cdot \Delta z) \quad (6)$$

Both a quadrupole component  $F = C_2 z^2$  and a dipole magnet component  $F = (2C_2 \cdot \Delta z) \cdot z$  are contained. Therefore, when measuring the field of a quadrupole, we must counteract the dipole component while reversing the quadrupole component [4], expressions [3] are:

$$s_2 = 1 - \beta_1^2 - \mu \rho^2 (1 - \beta_2^2) = 0 \quad (7)$$

$$s_1 = 1 + \beta_1 - \mu \rho (1 - \beta_2) = 0 \quad (8)$$

Other conditions to determine the radial sizes are as follows.

1. First we determine the outer radius  $r_1$  of the primary coil. Since the inner radius of the magnet is 40 mm, to make sure the outer radius of the coil is as close as possible to the designed good field edge, the primary coil outer radius is  $r_1 = 36$  mm, in consideration of the rotation gap.
2. Set the inner radius to outer radius proportional coefficient of the primary coil to  $\beta_1 = 0.75$ .
3. Set the turns ratio of the compensated coil to the primary coil to  $\mu = 1.5$ .

### Determination of Coil Turns

The higher harmonic amplitude is pretty small, usually the coil is wound more than 100 turns. A sufficient number of turns ensures that the signal-to-noise ratio is increased during the compensation measurement. However, the high resistance caused by excessive coil turns will vary with

ambient temperature, affecting the calibration and stability of the measurement system [5].

When deciding the number of turns, not only must the signal satisfy the demand of signal-to-noise ratio, but also the amplitude of the fundamental wave signal in the uncompensated measurement is smaller than the maximum range of the digital integrator input signal. *Metrolab's* high-precision integrator PDI5025 is used to process measurement results, which has a maximum range of 5V and a minimum range of 5mV [3].

Fundamental wave voltage signal induced by quadrupole field during uncompensated measurement is:

$$E_2 = M_{outer} L_{eff} r_1^2 B' (1 - \beta_1^2) \cdot \omega \quad (9)$$

Where  $L_{eff}$  is the effective length of the magnet,  $B'$  is the gradient,  $\omega$  is the rotating speed of the coil. To make  $E_2$  about 70~80% of the maximum range (5V) of the integrator, the turns number of the coil should be as small as possible to reduce the self-inductance and capacitance. The spinning speed of the coil  $\omega$  is 1r/2s.

### Determination of Coil Length

When the length of the coil is designed, the fringe field including both ends of the magnet should be taken into consideration, which can reduce the requirement for longitudinal positioning of the coil during the magnetic measurement [6]. The empirical formula is:

$$L_c \approx L_{Fe} + 4D \quad (10)$$

Where  $L_c$  is the coil length,  $L_{Fe}$  is the length of the magnet core,  $D$  is the working aperture of the magnet.

### Summary of Rotating Coil Parameters

When values of  $r_1$ ,  $\beta_1$  and  $\mu$  are respectively 36mm, 0.75, and 1.5, the calculation results include the four radii of the rotating coil and two turns number are shown in Table 2. The advantage of this scheme is that the sensitivity  $s_2$  is close to the ideal value of 0, indicating that the compensation is complete; the turns number of the coil is appropriate, the signal-to-noise ratio is high when measuring the multipole field, and the error caused by the change of the resistance with temperature is small; the four radii values are large which lower requirements for machining accuracy.

Table 2: Rotating Coil Parameters

Parameters	Value
$r_1$ (mm)	36.00
$r_2$ (mm)	25.49
$r_3$ (mm)	27.00
$r_4$ (mm)	16.50
$M_{outer}$	200
$M_{inner}$	300
$L_c$ (mm)	560
$s_2$	4.23E-5

## ERROR ANALYSIS

### Error Caused by Changes in Ambient Temperature

The coil material is G11, and the length of the coil changes as the ambient temperature changes. The thermal expansion coefficient of G11 material is  $\alpha = 1.7 \times 10^{-4} / (\text{°C} \cdot \text{m})$ .

In Eq.9, as the length of the coil is the length of the magnet core plus 4 times the working aperture of the magnet, the fringe field is already included, so the change of the length  $L_{eff}$  has no effect on the measurement results. In addition, the parameters  $M_{outer}$ ,  $B'$  and  $\omega$  are not affected by temperature. Therefore, the effect of temperature on  $E_2$  is

$$\frac{\Delta E_2}{E_2} = \frac{(r_1 + \Delta r_1)^2 [1 - (\beta_1 + \Delta \beta_1)^2] - r_1^2 \cdot (1 - \beta_1^2)}{r_1^2 \cdot (1 - \beta_1^2)} = \frac{[r_1(1 + \alpha \cdot \Delta T)]^2 \left\{ 1 - \left[ \frac{r_3(1 + \alpha \cdot \Delta T)}{r_1(1 + \alpha \cdot \Delta T)} \right]^2 \right\} - r_1^2 \cdot (1 - \beta_1^2)}{r_1^2 \cdot (1 - \beta_1^2)} \quad (11)$$

Where  $\Delta T$  is the temperature change, considering  $\alpha \cdot \Delta T$  is a small amount, its high order term is negligible. Simplify the above formula:

$$\frac{\Delta E_2}{E_2} = 2\alpha \cdot \Delta T \quad (12)$$

To make the error less than 1E-4, the temperature change  $\Delta T$  should be less than  $\pm 0.3$  °C. Considering that one measurement takes little time so the temperature is almost unchanged, it provides reference for multiple and repeated measurement.

### Sensitivity Error Analysis

The smaller the sensitivity coefficient is, the better the measurement accuracy is. Sensitivity error [3] is

$$(\Delta s_2)_{max} = 2\beta_1 |\Delta \beta_1| + 2\mu\rho^2 \beta_2 |\Delta \beta_2| + 2\mu\rho(1 - \beta_2^2) \Delta \rho \quad (13)$$

According to formula (13), we can get:

$$\Delta \beta_1 = \frac{\partial \beta_1}{\partial r_3} \Delta r_3 + \frac{\partial \beta_1}{\partial r_1} \Delta r_1 = \frac{1}{r_1} \Delta r_3 + r_3 \cdot \frac{1}{r_1^2} \cdot \Delta r_1 \quad (14)$$

$$\Delta \beta_2 = \frac{1}{r_2} \Delta r_4 + r_4 \cdot \frac{1}{r_2^2} \cdot \Delta r_2 \quad (15)$$

$$\Delta \rho = \frac{1}{r_1} \Delta r_2 + r_2 \cdot \frac{1}{r_1^2} \cdot \Delta r_1 \quad (16)$$

If the winding error of the coil is 0.01mm, according to the above formulae,  $\Delta \beta_1 = 4.86E - 4$ ,  $\Delta \beta_2 = 7.09E - 4$ ,  $\Delta \rho = 4.75E - 4$ ,  $(\Delta s_2)_{max} = 2.156E - 3$ . Bucking ratio, which is the reciprocal of sensitivity error and is used to estimate the compensation level, is 463.78.

If the winding error of the coil is 0.02mm, bucking ratio will reduce to half of the original and is 231.89, which weakens the compensation effect. It indicates that reducing the winding error of the coil is of great significance to improve the compensation level of the rotating coil.

## CONCLUSION

Quadrupoles are of great importance to the beamline, so the accuracy of quadrupoles are crucial. To measure the accuracy of quadrupoles, we employ the rotating coil measurement system which has advantage on measuring the integral field harmonics components. This article introduces the rotating coil measurement system at first, then focuses on the rotating coil parameter design and gives detailed analysis process and complete parameters for the L270 quadrupole including radial size, coil turns and coil length.

Besides, the errors analysis is carried out. We analyses the error caused by the change of ambient temperature and sensitivity error caused by the coil winding error. Together, they provide a reference on coil fabrication and calibration.

## REFERENCES

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